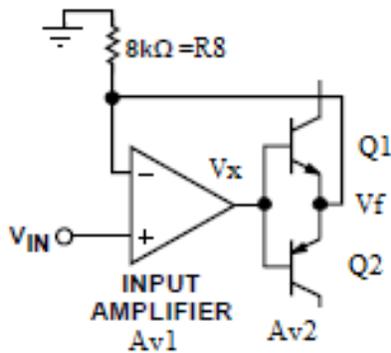


I will not be deriving the following two formulas as its very easy to do that,  
Voltage Gain of Emitter Follower,

$$A_v = \frac{\left( g_m + \frac{1}{r_e} \right)}{\left( g_m + \frac{1}{r_e} + \frac{1}{r_o} + \frac{1}{R_L} \right)}$$

Current gain of Common base,  $I_{out} = g_m V_{in}$

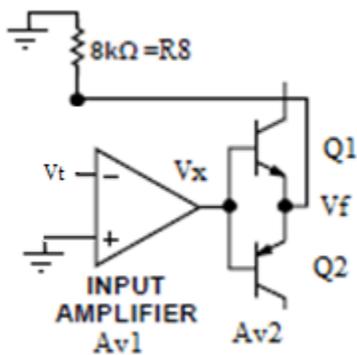
The AD746 ckt for the High Z AC coupled mode.



If Q1 operates, Q2 remains in cutoff and vice-versa. Av2 is same for both polarity of inputs and Q1=Q2.

The open loop gain without feedback is,  $A_{OL} = \frac{V_f}{V_x} \times \frac{V_x}{V_{in}} = A_{v1} A_{v2}$

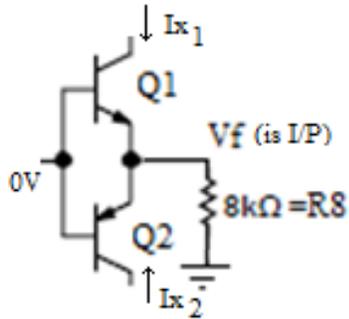
Q1 and Q2 acts as emitter follower for input at Vx.  $A_{v2} = \frac{\left( g_{m1} + \frac{1}{r_{e1}} \right)}{\left( g_{m1} + \frac{1}{r_{e1}} + \frac{1}{r_{o1}} + \frac{1}{R_L} \right)}$



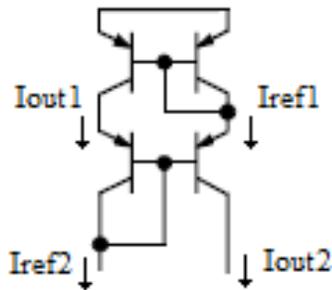
The loop gain =  $A\beta = \frac{V_f}{V_t} = -A_{v1} A_{v2}$

Thus the closed loop gain,  $\frac{V_f}{V_{in}} = A_{CL} = \frac{A_{OL}}{1 - A_{LG}} = \frac{A_{v1} A_{v2}}{1 + A_{v1} A_{v2}} \approx 1$  As  $A_{v1} \gg 1$

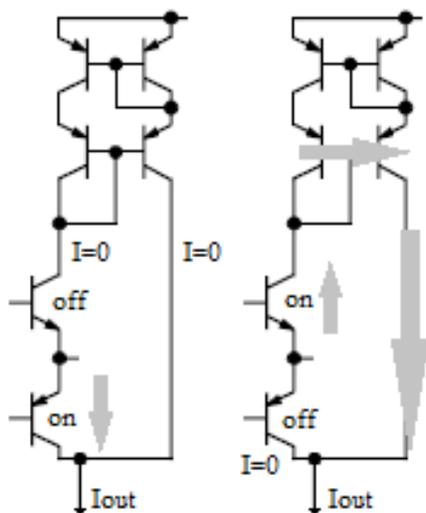
Thus Vf follows Vin and Vx remains at 0V.



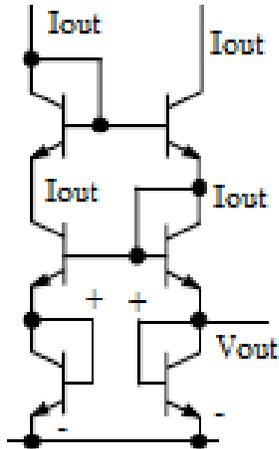
In this common base mode the output currents are,  
 When  $V_{in} = +|V_{in}|$ ,  $I_{x1} = 0$  and  $I_{x2} = -g_{m1}|V_{in}|$   
 When  $V_{in} = -|V_{in}|$ ,  $I_{x2} = 0$  and  $I_{x1} = g_{m1}|V_{in}|$



This is a bootstrapped current mirror. If we say  $\beta \gg 1$  and  $I_b$  is negligible, all the four currents are same. Like  $I_{ref1} = I_x$ ,  $I_{out1}$  will be  $I_x$  due to current mirroring. Applying KCL,  $I_{ref2}$  will also be  $I_x$ . This makes  $I_{out2} = I_x$  due to mirroring, which finally keeps  $I_{ref1}$  as  $I_x$  obeying KCL.



The current paths for two each cycle is shown in grey arrows. This arrangements gives rectified current  $I_{out} = g_{m1}|V_{in}|$



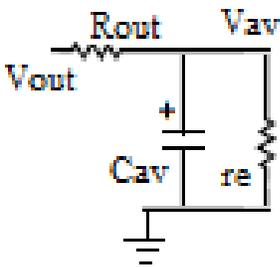
Yet another bootstrapped current mirror on top of two diode-connected loads. The input current in the loads are  $g_{m1}|V_{in}|$

Small signal analysis gives,  $V_{out} = \frac{I_{in}}{g_{m2} + \frac{1}{r_{e2}} + \frac{1}{r_{o2}}}$  where suffix 2 denotes the load transistor.

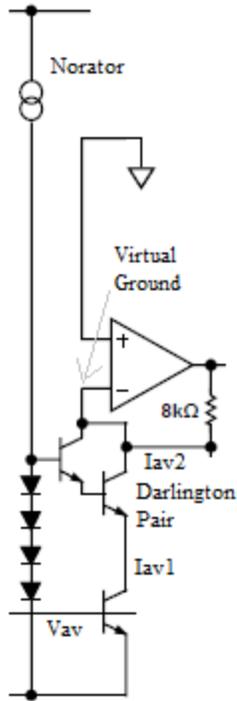
$$V_{out} = \frac{g_{m1}|V_{in}|}{g_{m2} + \frac{1}{r_{e2}} + \frac{1}{r_{o2}}}$$

Thus we get,

Low Pass filter (Averaging) filter,

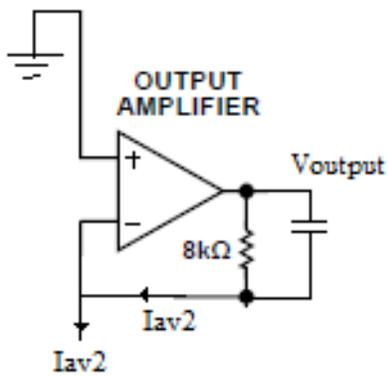


$V_{av}(s) = \frac{r_e}{r_e + R_{out} + sC_{av}r_eR_{out}} V_{out}(s)$  The ILT of this thing will be some integration which will do the average.  $r_e$  is emitter resistance of next stage and  $R_{out}$  is  $V_{out}/I_{in}$  given in previous paragraph.



Norator supports arbitrary voltage and current across it.

This stage is a common emitter followed by a Darlington pair.  $I_{av1} = g_{m5}V_{av}$  and  $I_{av2} = \beta_6^2 g_{m5}V_{av}$ . Here  $\beta_6$  is the beta of each transistor in Darlington pair and  $g_{m5}$  is transconductance of common emitter.



$$V_{output} = R_8 I_{av2} = \beta_6^2 g_{m5} R_8 \left[ AVG \left( \frac{g_{m1} |V_{in}|}{g_{m2} + \frac{1}{r_{o2}} + \frac{1}{r_{e2}}} \right) \right]$$

CF at output can be used for further smoothing of output.