

Electronic Devices and Circuits

Engineering Sciences 154

Laplace Transforms and Circuit Analysis

$$F(s) = \int_{0_-}^{\infty} f(t) \exp(-st) dt$$

Laplace transform as a means of signal representation

Example 1 - [unit step](#)

$$F(s) = \int_{\tau_0}^{\infty} \{1\} \exp(-st) dt = \left\{ -\frac{1}{s} \right\} \exp(-st) \Big|_{\tau_0}^{\infty} = \left\{ -\frac{1}{s} \right\} \exp(-s \tau_0)$$

Example 2 - [sinusoid](#)

$$\begin{aligned} F(s) &= \int_{0_-}^{\infty} \{ \cos(\omega t - \varphi) \} \exp(-st) dt && \Rightarrow \begin{cases} \cos(\omega t) & \text{when } \varphi = 0 \\ \sin(\omega t) & \text{when } \varphi = \frac{\pi}{2} \end{cases} \\ &= \int_{0_-}^{\infty} \frac{1}{2} \{ \exp[(j\omega - s)t - j\varphi] + \exp[-(j\omega + s)t + j\varphi] \} dt \\ &= \frac{1}{2} \left\{ \frac{1}{(j\omega - s)} \exp[(j\omega - s)t - j\varphi] \Big|_{0_-}^{\infty} + \frac{1}{-(j\omega + s)} \exp[-(j\omega + s)t + j\varphi] \Big|_{0_-}^{\infty} \right\} \\ &= \frac{1}{2} \left\{ \frac{-1}{(j\omega - s)} \exp(-j\varphi) + \frac{1}{(j\omega + s)} \exp(+j\varphi) \right\} \\ &= \begin{cases} \frac{1}{2} \left[\frac{-1}{(j\omega - s)} + \frac{1}{(j\omega + s)} \right] = \frac{s}{s^2 + \omega^2} & \text{for } \varphi = 0 \\ \frac{j}{2} \left[\frac{1}{(j\omega - s)} + \frac{1}{(j\omega + s)} \right] = \frac{\omega}{s^2 + \omega^2} & \text{for } \varphi = \frac{\pi}{2} \end{cases} \end{aligned}$$

Other Examples

Table of [Laplace transforms of common wave forms](#)

Table of [Laplace transforms of basic functions](#)

Table of [Laplace transforms of trigonometric functions](#)

Table of [Laplace transform pairs](#)

Laplace transform in circuit analysis

Property 1 - derivative

$$F^{deriv}(s) = \int_{0_-}^{\infty} \left\{ \frac{d}{dt} f(t) \right\} \exp(-st) dt$$

Integrate by parts

$$\begin{aligned} F^{deriv}(s) &= \int_{0_-}^{\infty} \frac{d}{dt} \{ f(t) \exp(-st) \} dt - \int_{0_-}^{\infty} (-s) f(t) \exp(-st) dt \\ &= s \int_{0_-}^{\infty} f(t) \exp(-st) dt - f(0_-) = s F(s) - f(0_-) \end{aligned}$$

Property 2 - integral

$$F^{integr}(s) = \int_{0_-}^{\infty} \left\{ \int_{0_-}^t \{ f(t') dt' \} \right\} \exp(-st) dt$$

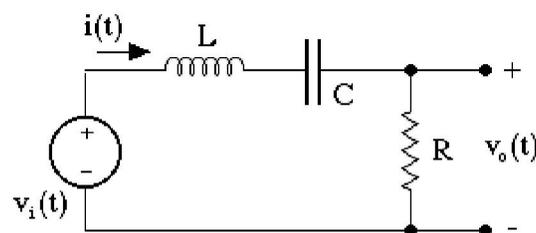
Integrate by parts

$$\begin{aligned} F^{integr}(s) &= \int_{0_-}^{\infty} \frac{d}{dt} \left\{ \left[\int_{0_-}^t \{ f(t') dt' \} \right] \left[\frac{\exp(-st)}{-s} \right] \right\} dt - \int_{0_-}^{\infty} \left[\frac{\exp(-st)}{-s} \right] \frac{d}{dt} \left\{ \left[\int_{0_-}^t \{ f(t') dt' \} \right] \right\} dt \\ &= \frac{1}{s} \int_{0_-}^{\infty} f(t) \exp(-st) dt = \frac{1}{s} F(s) \end{aligned}$$

Other Properties

Table of [Laplace transform properties](#)

RLC Circuits An Example of the Application of Laplace Transforms



Kirchoff's voltage law

$$v_i(t) = i(t)R + L \frac{di(t)}{dt} + \frac{1}{C} \int_{0_-}^t i(t') dt'$$

Kirchoff's voltage law transformed

$$\begin{aligned} V_1(s) &= R I(s) + s L I(s) - L i(0_-) + \frac{1}{s C} I(s) \\ &= \left[R + s L + \frac{1}{s C} \right] I(s) - L i(0_-) \\ &= Z(s) I(s) - L i(0_-) \end{aligned}$$

where $Z(s) = R + s L + \frac{1}{s C}$

Example 1 - A series RLC circuit excited by a unit impulse function - that is to say

$$V_1(s) = 1$$

(The unit impulse always gives the "natural response" of any circuit. Natural as compared to "forced" response.)

$$I(s) = \frac{1}{Z(s)} = \frac{1}{R + s L + \frac{1}{s C}} = \frac{1}{L} \left\{ \frac{s}{s^2 + s \frac{R}{L} + \frac{1}{LC}} \right\}$$

This is to be compared to the transform pair

$$f(t) = \sqrt{c^2 + d^2} \exp(-at) \cos\left(\omega t - \tan^{-1} \frac{d}{c}\right)$$

$$F(s) = \frac{c(s+a) + d\omega}{(s+a)^2 + \omega^2}$$

See the table of [Laplace transform pairs](#)

Therefore, if

$$a = \frac{R}{2L} \Rightarrow c = 1 \Rightarrow d = -\frac{\omega}{a} = -\frac{2\omega L}{R}$$

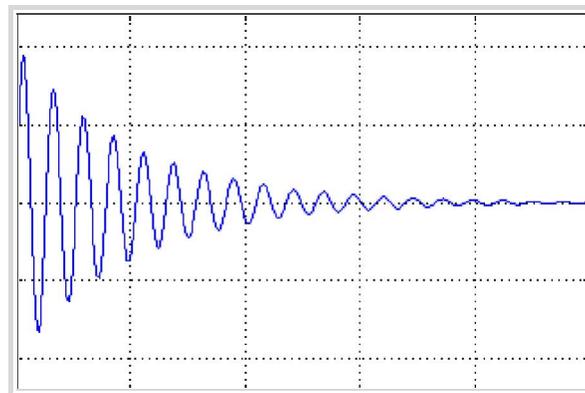
$$\omega^2 = \frac{1}{LC} - a^2 = \frac{1}{LC} - \left(\frac{R}{2L}\right)^2$$

$$\sqrt{c^2 + d^2} = \frac{\sqrt{a^2 + \omega^2}}{a} = \frac{2L}{R} \sqrt{\frac{1}{LC}}$$

$$\tan^{-1} \frac{d}{c} = -\tan^{-1} \left(\frac{2\omega L}{R}\right)$$

$$v_2(t) = i(t) R = 2 \sqrt{\frac{1}{LC}} \exp\left(-\frac{R}{2L}t\right) \cos\left\{\omega t + \tan^{-1}\left(\frac{2\omega L}{R}\right)\right\}$$

$$\text{where } \omega = \sqrt{\frac{1}{LC} - \left(\frac{R}{2L}\right)^2}$$



Comparison - A series RC circuit excited by a unit impulse function - that is to say

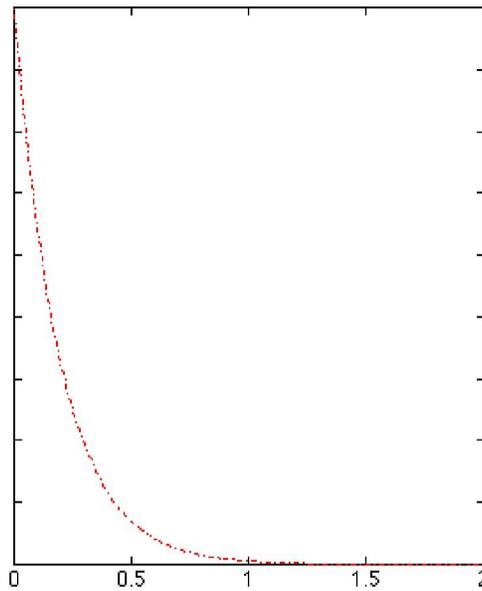
$$V_1(s) = 1$$

In general

$$\frac{V_o(s)}{V_1(s)} = \frac{RI(s)}{Z(s)I(s)} = \frac{sRC}{1+sRC}$$

and in particular

$$V_o(s) = \frac{sRC}{1+sRC} = s \frac{1}{s + \frac{1}{RC}}$$



Example 2 - A series RLC circuit excited by a unit cosine function - that is to say

$$V_1(s) = \frac{s}{s^2 + \omega^2}$$

$$I(s) = \frac{1}{L} \left\{ \frac{s}{s^2 + \omega^2} \right\} \left\{ \frac{s}{s^2 + s \frac{R}{L} + \frac{1}{LC}} \right\}$$

We need more methodology so go see [Partial Fraction Expansions](#)

See [Symbolic Inverse Laplace Transform Applet](#)

Last updated September 28, 2001